## Three Phase Bridgeless Interleaved Active Power Factor Correction Converter

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Abstract- In this paper, a new three-phase ac-dc bridgeless interleaved power factor correction topology is proposed for battery charging applications. The topology provides improved power factor and efficiency in comparison to conventional bridge rectifier topology, and leads to a decrease in charger size, charging time and hence, is cost effective. A detailed circuit operation for this topology is presented. Simulation results are included for a boost converter converting AC input voltage 230 V to 750 V DC output. Simulation results show an improved power factor and reduced THD, when compared to conventional topology. Power factor in proposed BLIL PFC is improved by 2.19% and THD is reduced by 26.25%.

Keywords: Bridgeless Interleaved, THD, Power Factor, Displacment factor, Distortion factor, MatLab.

#### 1. INTRODUCTION

A Universal battery charger is supplied using an external power supply. The battery charger module as shown in fig.1 consists of AC and DC filters in line and load side respectively, an AC-DC Power Factor Correction (PFC) Boost converter is used for power factor correction which is followed by a DC-DC converter. A controller is used to control the charger set up. The AC-DC PFC boost converter is an important part in charger module. This AC-DC has to be selected carefully. For this a three-phase Bridgeless Interleaved (BLIL) PFC boost converter is proposed which will reduce the THD and increases the power factor.



Fig.1. Block diagram of a universal battery charger

## 2. PROPOSED THREE-PHASE BLIL BOOST TOPOLOGY

The Three-phase BLIL PFC converter shown in Fig. 2 is proposed to overcome the problems in conventional converters. The proposed three-phase BLIL PFC Converter, consists of six MOSFETs, six diodes, six inductors Dr. V. Kamaraj<sup>3</sup> Professor Department of Electricals & Electronics SSN College of Engineering, Kalavakkam Tamilnadu, India kamarajv@ssn.edu.in<sup>2</sup>

and a capacitor parallel to load. A detailed converter operation a steady- state analysis is given in the following section



Fig.2. Proposed Three-Phase BLIL PFC Converter

# 3. CIRCUIT OPERATION AND STEADY STATE RIPPLE ANALYSIS

To analyze the circuit operation, the input line cycle has been separated into the positive and negative half cycles. Operation for each of the half-line cycles are explained in Sections 3.1 and 3.2 that follow.

## 3.1POSITIVE HALF CYCLE OPERATION

Referring to Fig. 2, during the positive half cycle, when Phase A and Phase B are conducting,  $Q_1/Q_2$  turn on and current flows through  $L_1$  and  $Q_1$  and continues through  $Q_2$  (and partially its body diode) and then  $L_2$ , returning to the line while storing energy in  $L_1$  and  $L_2$ . When  $Q_1/Q_2$  turn off, energy stored in  $L_1$  and  $L_2$  is released as current flows through  $D_1$ , through the load and returns through the body diode of  $Q_2$ back to the input mains. With interleaving, the same mode happens for  $Q_4/Q_5$ , but with a 180° phase delay. The operation for this mode is  $Q_4/Q_5$  on, storing energy in  $L_4/L_5$  through the path  $L_4$ - $Q_4$ - $Q_5$ - $L_5$  back to the input. When  $Q_4/Q_5$  turn off, en ergy is released through  $D_4$  to the load and returning through the body diode of  $Q_5$  back to the input mains.

During the positive half cycle, when Phase B and Phase C are conducting,  $Q_2/Q_3$  turn on and current flows through  $L_2$ and  $Q_2$  and continues through  $Q_3$  (and partially its body diode) and then  $L_3$ , returning to the line while storing energy in  $L_2$ and  $L_3$ . When  $Q_2/Q_3$  turn off, energy stored in  $L_2$  and  $L_3$  is released as current flows through D<sub>2</sub>, through the load and returns through the body diode of Q<sub>3</sub> back to the input mains. With interleaving, the same mode happens for  $Q_5/Q_6$ , but with a 180° phase delay. The operation for this mode is  $Q_5/Q_6$  on, storing energy in L<sub>5</sub>/L<sub>6</sub> through the path L<sub>5</sub>-Q<sub>5</sub>-Q<sub>6</sub>-L<sub>6</sub> back to the input. When  $Q_5/Q_6$  turn off, energy is released through  $D_5$ to the load and returning through the body diode of Q<sub>6</sub> back to the input mains. During the positive half cycle, when Phase C and Phase A are conducting,  $Q_3/Q_1$  turn on and current flows through L<sub>3</sub> and Q<sub>3</sub> and continues through Q<sub>1</sub> (and partially its body diode) and then L1, returning to the line while storing energy in  $L_3$  and  $L_1$ . When  $Q_3/Q_1$  turn off, energy stored in  $L_3$ and L<sub>1</sub> is released as current flows through D<sub>3</sub>, through the load and returns through the body diode of Q1 back to the input mains. With interleaving, the same mode happens for  $Q_6/Q_4$ , but with a 180° phase delay. The operation for this mode is  $Q_6/Q_4$  on, storing energy in  $L_6/L_4$  through the path  $L_6$ - $Q_6-Q_4-L_4$  back to the input. When  $Q_6/Q_4$  turn off, energy is released through  $D_6$  to the load and returning through the body diode of Q<sub>4</sub> back to the input mains.

#### 3.2NEGATIVE HALF CYCLE OPERATION

Negative half cycle operation is similar to that of the positive half cycle operation but the current flow direction will be in the opposite direction to that of the positive half cycle.

#### 4. FINDING VALUES OF L AND C

Designing the values of L and C is very important. In the proposed BLIL PFC Converter the values of L and C are designed as follows.

Let us consider the instant when Phase A and Phase B are conducting. During this interval the voltage equation is given as

$$V_{in}=(L1+L2)\frac{\Delta iL1}{(1-2)T_{b}}$$
(1)  

$$V_{0}-V_{in}=(L4+L5)\frac{\Delta iL4}{(1-D)T_{b}}$$
(2)

Assuming matched inductors,  $L_1$ ,  $L_2$ ,  $L_4$  and  $L_5$ , the input ripple current is the sum of currents in  $L_1/L_2$  and  $L_4/L_5$ 

$$\Delta \operatorname{Iin}-\frac{1}{L1+L2}\operatorname{Vo}(1-D)\operatorname{Ts}$$
(3)

From Equation 3

$$L1+L2=\frac{Ve(1-D)Is}{\Delta Iin}$$
(4)

$$\mathbf{L} = \frac{V_0(1-D)T_1}{2\Delta I in}$$
(5)

At this interval the change in capacitor voltage  $\Delta V c$  is given by

$$\Delta V c = \frac{1}{c} \int_{c0}^{c1} I c \, dt = \frac{1}{c} \int_{c0}^{c1} i L3 \, dt \qquad (6)$$

$$\Delta \mathbf{V} \mathbf{c} = \frac{\mathbf{i} \mathbf{L} \mathbf{3}}{\mathbf{c}} (\mathbf{t} \mathbf{1} \cdot \mathbf{t} \mathbf{0}) \tag{7}$$

$$\Delta V c = \frac{iL3}{c} (1 D) T \varepsilon$$
 (8)

From Equation 8

Where

$$C = \frac{iL3}{\Delta Vc} (1-D) Ts$$
 (9)

$$C = \frac{V_0^*(1-D)}{\Delta V_c^* f'^{\mathbb{R}}}$$
(10)

$$Ts = \frac{1}{f}$$
  
*tL*3 =  $\frac{V0}{R}$ 

R=output resistance

Equation 5 and 10 gives the value of L and C respectively.

#### 5. SIMULATION RESULTS

#### 5.1 CONVENTIONAL BOOST CONVERTER TOPOLOGY



Fig.3. MATLAB Implementation of Conventional Boost Converter Topology

Design parameters: Switching Frequency : 25 KHz Input Voltage : 230 V

Output Voltage	: 753 V
Duty Ratio	: 0.67
Inductor	:131mH
Resistor	: 200 Ω
Capacitor	: 760µF

Fig.4 and Fig.5 shows supply voltage and supply current and THD plot waveform of a Three-phase Conventional PFC respectively for duty ratio 67%



Fig.4 Supply Voltage and Supply Current waveforms



Fig.5 THD plot

## **Power Factor Calculation: Formulas used:** Power Factor= K<sub>d</sub> \* K<sub>p</sub>

 $K_d = \cos\varphi, K_p = 1/\sqrt{THD^2}$ 

 $\varphi = \text{Phase difference between input voltage and current}$   $\varphi = \text{Phase difference between input voltage and current}$  THD = Total Harmonic Distortion in input current  $K_d = \text{Displacement Factor}$  $K_p = \text{Distortion Factor}$ 

## 5.2 BRIDGELESS INTERLEAVED BOOST CONVERTER TOPOLOGY



Fig.6. MATLAB Implementation of Bridgeless Interleaved Boost Converter Topology

Design parameters:

Switching Frequency	: 25 KHz
Input Voltage	: 230 V
Output Voltage	: 757 V
Duty Ratio	: 0.67/0.33
L1,L2,L3,L4,L5,L6	: 66.7 mH
Capacitor C	: 975 µF
Resistor R	: 200Ω
R1,R2,R3,R,R5,R6	: 25mΩ

## DUTY RATIO: 67%

Fig.7 and Fig.8 shows supply voltage and supply current, THD plot and output voltage waveform of a Three-phase BLIL PFC respectively for duty ratio 67%







Fig.8 THD plot

### **Power Factor Calculation:**

 $\phi = 0.0022 \text{sec} = 24.3$   $Cos\phi = 0.9408$   $Kd = Cos\phi = 0.9408$  THD = 6.02% Kp = 0.9982 PowerFactor = Kp \* Kd = 0.9391

## DUTY RATIO: 33%

Design parameters:	
Switching Frequency	: 25 KHz
Input Voltage	: 230 V
Output Voltage	: 757 V
Duty Ratio	: 0.33
L1,L2,L3,L4,L5,L6	: 66.7mH
Capacitor C	: 975 µF
Resistor R	: 200Ω
R1,R2,R3,R,R5,R6	: 25mΩ

Fig.9 and Fig.10 shows supply voltage and supply current and THD plot waveform of a Three-phase BLIL PFC respectively for duty ratio 33%



Fig.9 Supply Voltage and Supply Current waveforms



Fig.10 THD plot

#### **Power Factor Calculation:**

 $\varphi = 0.0025 \text{sec} = 22.5$ Cos $\varphi = \text{Kd} = 0.9238$ THD = 18.63% Kp = 0.9830 PowerFactor = Kd \* Kp = 0.9080

## 6. COMPARISON OF THD AND POWER FACTOR

Since, it is three-phase we have considered duty ratio as 67%. It is observed that the THD is reduced and the Power Factor is improved in proposed three-phase BLIL topology compared to three-phase conventional topology

TOPOLOGY	THD (%)	POWER
		FACTOR
Three-phase conventiona	al 32.27	0.9172
(D=0.67)		
Three-phase BLI	L 6.02	0.9391
(D=0.67)		
Three-phase BLI	L 18.36	0.9080
(D=0.33)		

Table1. Comparison of THD and power factor

• Improvement in Power Factor =0.9391-0.9172=0.0219 =2.19%

## 7. CONCLUSION

Thus from the above simulation results the following two were inferred. Duty ratio of 67% is best suited for threephase converter topology. THD is reduced and Power Factor is improved in proposed three-phase BLIL PFC converter compared to Conventional topology.

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Reduction in THD =32.27-6.02=26.25%

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